CIGAL C++ sIGnal scAttering & waveLet fast representations

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Contents

| 1 | Introduction | 3 | | |
|--------------|--|-----------------|--|--|
| 2 | Audio File io 3 | | | |
| | 2.1 File Structure | 3 | | |
| | 2.2 Implementation \ldots | 5 | | |
| 3 | Fourier Transform | 5 | | |
| | 3.1 Definitions | 5 | | |
| | 3.2 Fast Fourier Transform | 6 | | |
| | 3.2.1 Danielson-Lanczos Lemma | 7 | | |
| | 3.2.2 Twiddle Factor Properties | 8 | | |
| | 3.2.3 Butterfly Scheme | 9 | | |
| | 3.3 Implementation | 9 | | |
| | 3.3.1 Inverse Fourier Transform | 11 | | |
| | 3.3.2 Graph | 11 | | |
| 4 | Spectrogram 11 | | | |
| - | 4.1 Algorithm | 12 | | |
| | 4.2 Implementation | 12 | | |
| | 4.2.1 Graph | 13 | | |
| | 4.3 Examples | 13 | | |
| 5 | Scattoring Notwork | 15 | | |
| 9 | 5.1 Introduction | 15 | | |
| | 5.2 Filter Bank implementation | 17 | | |
| | 5.2 Finter Dank implementation $\dots \dots \dots$ | 10 | | |
| | 5.2.1 Graph | 10 | | |
| | 5.3.1 Graph | 20 | | |
| | 5.4 Decomposition Implementation | 20 | | |
| | 5.5 Scattering Network Implementation | 21 | | |
| | 5.6 Examples | $\frac{21}{22}$ | | |
| c | | 0.4 | | |
| 0 | Data representation | 24 | | |
| | 0.1 Import data | 24 | | |
| | 0.2 Optimization | 24 | | |
| | 3.4 Implementation | 20 | | |
| |).4 Implementation | 20 | | |
| 7 | Validation on a large scale bioacoustic data set | 26 | | |
| | 7.1 Recall of previous attempts | 26 | | |
| | 7.2 Results with CIGAL | 26 | | |
| | .5 Perspective on BIRD2015 | 20 | | |
| 8 | Future work 28 | | | |
| \mathbf{A} | README 29 | | | |
| в | Licence GPL 30 | | | |
| | | | | |

1 Introduction

With the computational power available today, machine learning is becoming a very active field finding its applications in our everyday life. One of its biggest challenge is the classification task involving data representation (the preprocessing part in a machine learning algorithm). In fact, classify linearly separable data is easily done. The aim of the preprocessing part is to obtain well represented data by mapping raw data into a feature space where simple classifiers can be used efficiently. For example, everything around audio processing uses MFCC until now. This toolbox gives the basic tools for audio representation using the C++ programming language by providing an implementation of the Scattering Network [4] which brings a new and powerful solution for these tasks. The tool-kit of reference in scattering analysis is the SCATNET from Mallat et al.¹. This tool is an attempt to have some of the scatnet features more tractable in large dataset. Furthermore, the use of this toolbox is not limited to machine learning preprocessing. It can also be used for more advanced biological analysis such as animal communication behaviours analysis or any biological study related to signal analysis. One motivation for this work is the collaboration between DI ENS and the university of Toulon through the SABIOB Scaled Acoustic project. [15] [14]. This toolbox gives out of the box executables that can be used by simple bash commands. Examples are given in the README file. Finally, for each presented algorithm, a graph is provided in order to summarize how the computation is done in this toolbox.

2 Audio File io

2.1 File Structure

The WAV file is an instance of a Resource Interchange File Format (RIFF) defined by IBM and Microsoft. The header part of this file is made of complementary chunks describing the architecture of the wav allowing easy information storing. Let's see how these chunks are organized in a WAV file :

¹http://www.di.ens.fr/data/software/scatnet/



Figure 1: The Canonical WAV file format https://ccrma.stanford.edu/ courses/422/projects/WaveFormat/

The file is made of three main chunks each having a specific role that we will describe here.

- ChunkID identifies the type of the first chunk with four characters : "RIFF".
- ChunkSize is the size of the file left from this point. It will be 36 (sum of the other chunks sizes) plus Subchunk2Size and this can be easily seen by summing the different sizes on the right of the header representation.
- Format will be four characters : "WAVE" (this allows us to check if we are really reading a wav file during the process).
- SubChunk1ID identifies the second chunk. It is a four characters name : "fmt " and starts the data description block.
- SubChunk1Size is simply the size of this block which is 16.
- AudioFormat, also called Format tag, is the option indicating the data compression used. It is almost always equal to 1 which stands for : no compression is used.
- NumChannels is the number of channels (1 for mono and 2 for stereo).
- SampleRate is simply the number of samples per second, the frequency.

- ByteRate is the average number of bytes per second, this can be found with the following formula : SampleRate * NumChannels * $\frac{\text{BitsPerSample}}{8}$.
- BlockAlign won't be necessary for us. It can be computed with the formula : $NumChannels * \frac{\text{BitsPerSample}}{8}$.
- BitsPerSample can either be 8 or 16 but in general the later is used.
- SubChunk2ID identifies the last chunk block, it is made of the four characters : "data".
- Subchunk2Size is the size of the file left which is just the size of the data.
- Data is the block containing the values of the signal in the standard pulsecode modulation representation.

2.2 Implementation

The use of the built-in class WAV is simple, the only thing to provide is the name of the wav file. This can be done during the instantiation of the class or at any other time. Let's look at an example.

```
WAV<br/>double> Signal("mysignal.wav");<br/>WAV<br/>double> Signal2;<br/>"mysignal.wav">>Signal2;<br/>"mysignal2.wav">>Signal;
```

The template parameter can be ignored leading to the default value : float. One instance of the class can be used for different wav files which can be useful. This WAV variable allows easy interactions and can provide informations about the loaded file :

```
Signal._Size;
Signal._SampleRate;
Signal.NumberOfChannel;
Signal.NimberOfChannel;
```

Finally, to export the loaded file two options are available. Firstly, it is possible to export it into a txt file which only export the signal data disregarding all the other informations. This loss can be avoided using a special method which exports the data back into a wav file, with the following syntax :

```
Signal>>" newsignal.txt"; //to .txt
Signal.PrintWav("newsignal.wav"); //to .wav
```

With this implementation, WAV can be seen as a special type. Note that no normalization is used. In fact, only the user can define the normalizing constant he is interested in (max, L^2 -norm,...) and so has to apply it after the import. Finally, for an external use of this toolbox, one should not need to use this class since it is just here as a input/output convenience for the other algorithm we will now describe.

3 Fourier Transform

3.1 Definitions

A sinusoidal wave is characterised by three parameters: amplitude, frequency and phase.

- The amplitude is the amount the function varies, positively or negatively, from zero in the y direction.
- The frequency is how many complete cycles there are of the wave in unit distance on the x axis (which often measures time).
- The phase is relevant when comparing two waves of the same frequency. It is how much (measured in degrees or radians) one is shifted relative to the other on the x axis.

This terminology comes from sound engineering where higher frequency sounds have higher pitch and waves of greater amplitude are louder. As an alternative of specifying the frequency, the number of cycles in unit distance, we can instead specify the wavelength : the length of one cycle. The higher the frequency, the shorter the wavelength. The lower the frequency the longer the wavelength. The Nyquist frequency is the maximum frequency that can be detected for a given sampling rate and it is half of it. This is because in order to measure

given sampling rate and it is half of it. This is because in order to measure a wave one needs at least two sample points to identify it (trough and peak). We will abbreviate the continuous Fourier transform with CFT and the discrete Fourier transform with DFT.

Interpretation of the CFT Using the Euler's formula, we can see the Fourier Transform as a decomposition of a signal into complex sinus by the use of convolutions.

$$e^{ix} = \cos(x) + i\sin(x)$$

$$\begin{split} \hat{f}(\xi) &= \int_{-\infty}^{\infty} f(x)e^{-2\pi i x\xi} dx \\ &= \int_{-\infty}^{\infty} f(x)(\cos(-2\pi x\xi) + i\sin(-2\pi x\xi)) dx \\ &= \int_{-\infty}^{\infty} f(x)\cos(-2\pi x\xi) + f(x)i\sin(-2\pi x\xi) dx \\ &= \int_{-\infty}^{\infty} f(x)\cos(-2\pi x\xi) dx + \int_{-\infty}^{\infty} f(x)i\sin(-2\pi x\xi) dx \end{split}$$

3.2 Fast Fourier Transform

We will now denote x_k as the k^{th} value on the signal in the time space and X_k the k^{th} value of the signal in the frequency domain, N will denote the length of the signal. A length of N means the indices range from 0 to N - 1. The fact Fourier transform (FFT) is a justance of DFT which is able to perform

The fast Fourier transform (FFT) is a instance of DFT which is able to perform the DFT in $O(N \log(N))$ complexity.

The DFT formula using the Twiddle Factor notation :

$$\forall k \in \mathbb{Z}, X_k = \sum_{n=0}^{N-1} x_n e^{\frac{-2\pi i k n}{N}}$$
$$X_k = \sum_{n=0}^{N-1} x_n W_N^{kn}$$

As we can see, we need to perform N operations for each $X_k, k \in \{0, 1, ..., N-1\}$ thus we are in $O(N^2)$ complexity.

Note that it is possible to use the scaling factor $1/\sqrt{N}$ in order to have an unitary operator (Parseval's theorem) which implies that the sum (or integral) of the square of the function is equal to the sum (or integral) of the square of its transform which is not needed in this toolbox thus not used.

In order to go from N^2 operations to $N\log(N)$ operations, three main concepts have to be defined :

- The Danielson-Lanczos Lemma
- The Twiddle Factor properties
- The Butterfly Scheme

3.2.1 Danielson-Lanczos Lemma

This theorem is the foundation of the FFT by allowing a divide and conquer strategy. In fact, we have the following relations :

$$X_{k} = \sum_{n=0}^{N-1} x_{n} e^{\frac{-i2\pi kn}{N}}$$

$$= \sum_{n=0}^{\frac{N}{2}-1} x_{2n} e^{\frac{-i2\pi 2kn}{N}} + x_{2n+1} e^{\frac{-i2\pi (2k+1)n}{N}}$$

$$= \sum_{n=0}^{\frac{N}{2}-1} x_{2n} e^{\frac{-i2\pi kn}{N/2}} + \sum_{n=0}^{\frac{N}{2}-1} x_{2n+1} e^{\frac{-i2\pi 2kn - i2\pi n}{N}}$$

$$= \sum_{n=0}^{\frac{N}{2}-1} x_{2n} e^{\frac{-i2\pi kn}{N/2}} + W_{N}^{k} \sum_{n=0}^{\frac{N}{2}-1} x_{2n+1} e^{\frac{-i2\pi kn}{N/2}}$$

For every X_i we can now divide the N sums into two different summation group (Even and Odd). Note that for the special case N = 2 the sums are removed and n is replaced by 0 which means that we are left with a simple linear combination of the input signal and the Twiddle Factor. If we apply this recursively we obtain the following architecture :



And now for any given input size we are able to break it done into a linear combination of the input signal with twiddle factors. For example, if N = 4 we have after full decomposition :

$$X_k = x_0 + W_2^k x_2 + W_4^k x_2 + W_4^k W_2^k x_3$$

And for N = 8:

$$X_{k} = x_{0} + W_{2}^{k} x_{4} + W_{4}^{k} x_{2} + W_{4}^{k} W_{2}^{k} x_{6} + W_{8}^{k} x_{1} + W_{8}^{k} W_{2}^{k} x_{5} + W_{8}^{k} W_{4}^{k} x_{3} + W_{8}^{k} W_{4}^{k} W_{2}^{k} x_{7} + W_{8}^{k} W_{4}^{k} x_{7} + W_{8}^{k} W_{8}^{k} + W_{8}^{k} X_{7} + W_{8}^{k} W_{8}^{k} + W_{8}^{k} X_{7} + W_{8}^{k} W_{8}^{k} + W_{8}^{k} +$$

This puts a constraint though, the signal length has to be a power of 2. The number of decomposition is thus $\log_2(N)$. If the signal size is not a power of 2 it is necessary to use zero padding (add as may 0 as necessary at the end of the input). Padding with 0 in time domain leads to an interpolation of the FFT. Middle zero padding the FFT (in the frequency domain) interpolates the IFFT (time domain). Periodizing in the frequency domain implies sub-sampling in the time domain (this will be useful for the Scattering Network).

One last thing to notice here is the order of the input values in the decomposition. Because of the nature of this decomposition (even/odd) we end up with the x terms being rearranged in a specific order : the bit-reversal order. This can be found by taking the symmetric of the binary position of the input value as seen in this little example for N = 8:

$$\begin{array}{c} 0:000 \rightarrow 000:0\\ 1:001 \rightarrow 100:4\\ 2:010 \rightarrow 010:2\\ 3:011 \rightarrow 110:6\\ 4:100 \rightarrow 001:1\\ 5:101 \rightarrow 101:5\\ 6:110 \rightarrow 011:3\\ 7:111 \rightarrow 111:7 \end{array}$$

3.2.2 Twiddle Factor Properties

Complexity has already been broken down but we can still optimize the implementation by exploiting the Twiddle Factor properties using roots of unity properties. In fact we have :

$$W_N^k = e^{\frac{-i2\pi k}{N}} = \cos(2\pi k/N) - i\sin(2\pi k/N)$$

Thus for N = 2:

$$W_2^0 = W_2^2 = W_2^4 = \dots$$
$$W_2^1 = W_2^3 = W_2^5 = \dots$$

And for N = 4

$$\begin{split} W^0_4 &= W^4_4 = W^8_4 = \dots \\ W^1_4 &= W^5_4 = W^9_4 = \dots \\ W^2_4 &= W^6_4 = W^{10}_4 = \dots \\ W^3_4 &= W^7_4 = W^{11}_4 = \dots \end{split}$$

And so on using trigonometric properties, with functions here being $N\pi$ -periodic. So using this will allow us to perform less Twiddle Factor computation at each level.

3.2.3 Butterfly Scheme

Finally, the last brick is the butterfly scheme that can be seen in the next diagram 3.3 allowing an in-place FFT which is memory friendly.

3.3 Implementation

Firstly, our implementation is made of three nested loops, the main one which will go through the $\log(N)$ levels of decomposition. The second one will go through the blocks inside a specific level (the last level as 1 block whereas the first level as N/2 blocks). Finally the last loop will go inside a block (a block on the first decomposition level will have size 2 while the block in the last decomposition level will be of size N). For each level (i), only 2^i Twiddle factors are computed in the main loop where i is the decomposition level from 0 to $\log(N) - 1$. A simple temporary variable is used in order to perform the swapping operations. Here is an instance of this implementation for N = 8 and a human friendly output explaining the performed steps.



An 8 Input Butterfly. Note, you double a 4 input butterfly, extend output lines, then connect the upper and lower butterflies together with diagonal lines.

Figure 2: Full FFT computation with N = 8 [1]

| | | Level : 2 |
|-----------------------------|--------------------------------|--------------------------------|
| Level : 0 | Level : 1 | W[0] = W(0, 8), W[1] = W(1, 8) |
| W[0] = W(0, 2) | W[0] = W(0, 4), W[1] = W(1, 4) | W[2] = W(2, 8), W[3] = W(3, 8) |
| Block : 0 | Block : 0 | Block : 0 |
| signal[1] * = W[0] | signal[2] * = W[0] | signal[4] * = W[0] |
| tmp=signal[0] | tmp=signal[0] | tmp=signal[0] |
| signal[0] + = signal[1] | signal[0] + = signal[2] | signal[0] + = signal[4] |
| signal[1] = tmp - signal[1] | signal[2] = tmp - signal[2] | signal[4] = tmp - signal[4] |
| | signal[3] * = W[1] | signal[5] * = W[1] |
| Block : 1 | tmp=signal [1] | tmp=signal[1] |
| signal[3] * = W[0] | signal[1] + = signal[3] | signal[1] + = signal[5] |
| tmp=signal [2] | signal[3] = tmp - signal[3] | signal[5] = tmp - signal[5] |
| signal[2] + = signal[3] | | signal[6] * = W[2] |
| signal[3]=tmp-signal[3] | Block : 1 | tmp=signal [2] |
| | s i g n a l [6] * = W[0] | signal[2] + = signal[6] |
| Block : 2 | tmp=signal [4] | signal[6] = tmp - signal[6] |
| signal[5] * = W[0] | signal[4] + = signal[6] | signal[7] * = W[3] |
| tmp=signal[4] | signal[6] = tmp - signal[6] | tmp=signal[3] |
| signal[4] + = signal[5] | signal[7] * = W[1] | signal[3]+=signal[7] |
| signal [5] = tmp-signal [5] | tmp=signal [5] | signal[7] = tmp - signal[7] |
| | signal[5]+=signal[7] | |
| Block : 3 | signal[7] = tmp - signal[7] | |
| signal[7] * = W[0] | | |
| tmp=signal[6] | | |
| signal[6]+=signal[7] | | |
| signal [7] = tmp-signal [7] | | |

The Twiddle Factors are computed at the start of each main loop computing the needed values which are then reused throughout the blocks, meaning that for the first level only one value is computed and then reused all along the blocks. Here is an example of the use :

```
WAV> wav("signal.wav"); //load a wav into float type array
fft > signalfft(signal.ptr(), signal._Size); //default padding option=1
signalfft.ComputeFFT();
signalfft.ComputeFFT(); //get back to the original signal
signalfft[2]; //access the second coefficient
signalfft]>" signalfft.txt"; //export it
wav<<"processed.wav"; //load a new wav
signalfft.ComputeFFT((wav.ptr(),wav._Size); //perform a new FFT
```

Note that the parameters of the fft class are by default float and float, the first one stands for the type of the input signal and the latter for the coefficients type (complex<float>). Finally the padding option which by default is 1 can be set to 0 if the user is sure that the given signal is already a power of 2, this force to skip the padding part resulting in faster computation. Also the coefficients are stored as complex type even after having performed an IFFT meaning that one needs to use a typecast to retrieve the original float type signal for example.

3.3.1 Inverse Fourier Transform

In order to simplify the algorithm we sill use the following formula :

$$IFFT(x) = \frac{1}{N} conj(FFT(conj(x)))$$

3.3.2 Graph



Figure 3: FFT Summary Diagram

4 Spectrogram

Each X_k is a complex number that encodes how strongly the oscillation at this frequency is represented in the data but by doing an FFT we loose the time component. A useful tool is the spectrogram allowing to retrieve part of the time information. The main idea is to perform multiples FFT on a signal each one being located enough in time so the frequency information gained by the FFT can also be linked to a more or less specific time position in the signal. Note however that precision in both time and frequency is impossible to get but depending on the needs one can choose which one to enhance by modifying the size of the considered window. Larger window gives better frequency resolution but lesser time precision and vice-versa. It is easy to picture the fact that smaller windows are better for the high frequency part allowing good time precision while for low frequency a larger window has to be used for being able to capture it. This problem is lessen in wavelet decomposition and thus the scattering network since this window size is not constant anymore.

4.1 Algorithm

Conceptually a spectrogram is computed with the following scheme :

- splitting the signal into overlapping (or not) parts of equal length defined by the user.
- applying to each of these chunk a windowing function (typically hanning or hamming) in order to remove artefacts by periodizing the function so the limit points (start and end of the chunk) are equal. This part is called apodization
- computing the FFT on each of these chunks
- for each computed FFT, taking the absolute value of the coefficients will give the columns of the spectrogram.

The windowing is needed since the FFT computation presumes that the input data repeats over and over. This is important when the initial and final values of the data are not the same because the discontinuity causes artefacts in the spectrum computed by the FFT.

In addition, in this toolbox, only the first half of the FFT coefficient are put into the spectrogram thus avoiding symmetrical redundancy. This is due to the fact that our input signal is real and so the second half of the FFT coefficients is simply the complex conjugate of the first half, since in the spectrogram we display the absolute value of the coefficients, we get symmetry about the middle point.

Most window functions afford more influence to the data at the center of the window than tohe t data at the edges, which represents a loss of information. To mitigate that loss, it is common to use overlapping in time (usually 50%).

4.2 Implementation

It is important to note that the spectrogram (2D-matrix) is stored by column and not by line for faster computation. In fact, during the spectrogram calculation we need to access this matrix column-wise. The operator [] returns the column while the operator () takes two arguments and return the corresponding value in a normal way. Let's look at an example :

spectrogram <> b("signal1.wav",256,0.25);//default window function : hamming WAV<> wav("signal2.wav");//load another wav b.Perform(wav.ptr(),wav._Size);//compute spectrogram given these new entries //and default parameters with the already declared spectrogram variable b>>"lifespectro.txt";//write the matrix into a txt file b[1][0];//second column, first element b(0,1);//first line second element same result as above

The template parameter defines the coefficients type. The default value is float. Also note that no transformation is performed after the absolute value is computed, which means that if one want to apply a logarithmic function (most common one) this has to be done after computation.

The apodization can be done using one of the available windowing function :

- hamming
- hanning
- triangular
- $\bullet\,$ hann poisson

but can also be used with a specific user defined function passed as last argument when calling the Perform method.

4.2.1 Graph



Figure 4: Spectrogram Summary Diagram

4.3 Examples

Let's look at some spectrogram examples. Note that a logarithmic function has been applied to the computed values (improving coefficient representation for us). The signals are from a bird of the BIRDLIFE CLEF Challenge 2014 2 . and a Inia dolphin.

²http://www.imageclef.org/lifeclef/2015



Figure 5: Inia Dolphin Slow Clicks : Spectrogram 128~50%



Figure 6: Inia Dolphin Fast Clicks : Spectrogram 12850%



Figure 7: Bird : Spectrogram 512, 50%

5 Scattering Network

The Scattering Network aims to find a better data representation after numerous transformations of a raw input. It's been developed by Stéphane MALLAT and its team and his team in matlab, which is not the fastest implementation. In fact, this algorithm just started to be applied in concrete challenges and problems. We will review its core ideas and the implementation architecture I chose.

5.1 Introduction

The basic idea is to perform series of linear and non linear operations. The linear operations are done through the convolutions while the non linear ones are the use of the absolute value on these convolutions. The use of the latter allows fast convergence by the contractive property. The convolutions are basically decomposing the signal into a wavelet basis. A parallel can be made with the FFT and the complex sine decomposition. The structure itself of the network can be compared to a Convolutionnal Neural Network where the filters are computed and fully determined by the meta parameters while in a CNN they are learned during training. This is a huge difference in term of computation time allowing good representation without training. We have to keep in mind that filters generation is also complex and time consuming.

The mapped data into the feature space can be used for simple data analysis or data learning but it finds its best use in classification. In fact, this feature space is much more suited for the use of linear classifier. Note that in this implementation we won't look at the reconstruction problems since our main goal is not to use the Scattering Network for compression, reconstruction,... Let's look at the general picture of the scattering network and analyse it briefly.



Figure 8: Scattering Summary.[6]

In this case the scattering network is made of 3 layers. Each layer has lowpass filters (ϕ) and high-pass filters (ψ). In our specific case of 1D signals, there is only one ϕ per layer. Given an input signal x of size N we perform a low-pass decomposition (S_0x) by performing the convolution $x \star \phi$ and a high-decomposition leading to a output size of $2N/T \times$ NumberOfPsisFilters by performing NumberOfPsisFilters convolutions $x \star \psi_{i,\lambda 1}$ where $\psi_{i,\lambda 1}$ is the i^{th} filter of the ψ -filter bank generated by the meta parameters $\lambda 1$. Finally on this high-decomposition is applied the absolute value operation.

Then for the second layer, each one of the previous high-decomposition is treated as an input signal and the same algorithm is performed. Details about this will be given in the scattering layer section 5.3 but we can already note that the meta parameters are specific to a scattering layer Finally let's review what the meta parameters are about :

- T determines the time resolution by changing the size of the filter. Small T is suited for important time precision for high frequency signals.
- Q determines the quality factor (the number of filters per octave)
- J determines the number of octave to go through.
- PE (Periodization Extent) constant used in the filter periodization (1 by default)

In order to respect this architecture, this toolbox uses a specific structure : MetaParam using default parameters and a TtoJ method :

Let's now see the details of each implementation level and emphasize the implementation architecture used.

5.2 Filter Bank implementation

Filters are created through the constructor of the Filter1D class. Given metaparameters and a support size, the constructor will initialize all the wanted variables and compute the actual filters. Note that the Filter1D class has two children : the MorletFilter1D and GaborFilter1D. These two specializations have their own filter generation algorithm. This also means that if one wants to implement a new filter, the only thing to do is to create another class of the name of this filter, inherit from the Filter1D class and implement the coefficients generation method.

Note that the constructor can be used in two different ways :

- Giving support size, meta parameters, and the position of the filter in this configuration (gamma)
- Giving a support size, a σ and a ξ .

The first one is more practical for the ψ generation since the size and the meta parameters are fixed for a layer, we just have to loop through γ (the filter number in the filter bank). On the other hand, the second constructor is simpler for the ϕ filter generation, in fact, since only one low-pass filter is made per layer, we just have to compute ξ and σ for this filter.

Here is an example with arbitrary coefficients :

```
Filter1D* BankFilter=new Filter1D[5];
BankFilter[0] = GaborFilter1D(500,0,1,2);//500 points, xi=0,sigma=1,PE=2
for (int i=1;i<5;++i)
BankFilter[i]=MorletFilter1D(500,2+0.5*i,0.2*i);//500 points, xi=f(i),
//sigma=g(i),PE=1 (default)
ofstream file("filters.txt");
for (int i=0;i<5;++i){
file <<BankFilter[i];// use of the overloaded operator
file <<"\n";
}
delete[] BankFilter;
file.close();
```

Giving the following result :



Figure 9: Filters generation example, orange : Gabor filter, blue : Morlet wavelets.

The filters are directly computed in the Fourier domain to speed up the decomposition algorithm, indeed we only have to compute the FFT of the input to perform the decomposition algorithm now. Here ξ corresponds to the central frequency and so to the global maximum position. It can be seen as a position parameter while σ is a scale parameter. In practice, in order to generate the filters we always take the mother coefficients that are transformed through a scale coefficient following exponential change. We have then as mother coefficients :

$$\Xi = \frac{\pi}{2} * (2^{-1/Q} + 1)$$
$$\Sigma = \sqrt{3} * (1 - 2^{-1/Q})$$

The scaling factor for the filter i is $:\lambda_i = 2^{-i/Q}$ which leads to the following coefficients for any given filter i for a specific layer having the same meta parameters .

$$\xi_i = \Xi * \lambda_i$$
$$\sigma_i = \Sigma * \lambda_i$$

Filters In this implementation, high-pass filters are Morlet wavelets while lowpass filters are Gabor filters. Note that Morlet filters are actually another name for Gabor kernels. The difference between the Gabor function (non-zero-mean function) and the Gabor kernel (zero-mean function) is that the Gabor kernel satisfies the admissibility condition for wavelets (integral equals to 0), thus being suited for multi-resolution analysis. The admissibility condition ensures that the inverse transform and Parseval formula are applicable.

Filter Periodization In order to increase resolution of the filters, we can compute them on a bigger interval than the one we are interested in and then periodize them in the Fourier domain :

$$f(x) = \sum_{n \in \mathcal{Z}} f(x + 2\pi n)$$

In practice nothing assures the convergence for any function f but our filters are generated through Gaussian functions which assure convergence. In practice, we use $n \in \{-PE, -PE + 1, ..., PE, PE + 1\}$ with $x \in \{x \in \mathbb{R}, i = 0, ..., T - 1 : x = i * 2\pi/T\}$ which is similar to $x \in \{0, 2\pi/T, 2 * 2\pi/T, ..., (T - 1)2\pi/T\}$. With this definition x covers $[0, 2\pi[$ with T points linearly separated by a distance of 1/T. In all the examples presented here a periodization extent of 1 is used. The n coefficients affect the range on which the wavelet is evaluated which grows with bigger n:

$$[-2\pi * PE, 2\pi * (1 + PE) - 1/T]$$

It is then shrunk into the desired support size by the periodization process.

5.2.1 Graph



Figure 10: Filter1D Summary Diagram

5.3 Layer Implementation

The role of this class is to be the link between the raw input, the meta parameters, and the bank filters by performing the decomposition process. Firstly, this class takes a 2D input (the input signal has to be transformed for the first layer). This allows an easy link between layers by directly setting the input of the next one as the output of the previous one.

Given the input, private variables are computed determining the structure of the class by computing variables that will be passed to the next layer such as the size of the output (given the input size and the number of ψ filters :Q * J). Then when all the ψ filters are available a Littlewood-Paley normalization is performed (due to the logarithmic spaced filters). After this, the filters are generated using the Filter1D class. The Decomposition can now be performed.

Note that the decomposition is stored as a 2D matrix for every layer. Normally, layer *i* has a dimension of i + 1 which is not true in this toolbox. In fact, in this toolbox, the graph structure of the scattering network5.1 has been kept through 2D matrices. For example for the second layer if we have k filters for L1 and l filters for L2 and with $\psi_{i,j}$ being the h^{th} filter of the i^{th} layer, the

high-decomposition matrix of L2 will be :

$$\left(\begin{array}{c} ||x \star \psi_{1,1}| \star \psi_{2,1}| \\ \dots \\ ||x \star \psi_{1,1}| \star \psi_{2,l}| \\ ||x \star \psi_{1,2}| \star \psi_{2,1}| \\ \dots \\ ||x \star \psi_{1,2}| \star \psi_{2,l}| \\ \dots \\ ||x \star \psi_{1,k}| \star \psi_{2,1}| \\ \dots \\ ||x \star \psi_{1,k}| \star \psi_{2,l}| \end{array} \right)$$

If one wants to select a specific $\psi_{2,i}$ decomposition it can be done by selecting the lines $i, i + l, i + 2l, \ldots$ and thus only analysing one path of the scattering network.

5.3.1 Graph



Figure 11: Scattering Layer Summary Diagram

5.4 Decomposition Implementation

The core of the algorithm lies in this decomposition. Firstly, the convolution defined in the section 5.1 is redundant and so is only performed on every T/2 spaced points. This implies a reduced output length and faster computation. Thus, it is necessary to perform a periodization before computing the IFFT (allowing a time sub-sampling). The output length must then be InputSize*2/T. Doing this for each psis filter gives the output of the layer. Here is a simple scheme to emphasize the algorithm :

Algorithm 1: Decomposition Algorithm

With "periodize" being the function that will periodize the result in order to sub-sample in the time domain to obtain the desired output size. In the algorithm, after a layer has performed the decompositions, filters are freed in order to reduce memory consumption. In fact, filters of computed layers wont be reused and it would be a waste to keep them.

5.5 Scattering Network Implementation

Finally here is how to perform the Scattering Network on a signal and to save the outputs :

```
MetaParam* opt=new MetaParam[3];
opt[0]=MetaParam(8,30,4,1);
opt[1]=MetaParam(64,1,1,1);
opt[2]=MetaParam(1024,1,1,1);
ScatteringNetwork decomposition("signal.wav",opt,3);
ofstream file;
file.open("layer1.txt");
file <cdecomposition[0];
file <cdecomposition[0];
file <cdecomposition[1];
file <cdecomposition[1];
file <cdecomposition[1];
file <cdecomposition[2];
file <cdecomposition[2];
file <cdecomposition[2];
file <cdecomposition[2];
file <cdecomposition[2];
file <cdecomposition[2];
file <cdecomposition[2];</pre>
```

In fact the operator [] is overloaded to return the specific layer which itself uses its overloaded operator to export the coefficients.



Figure 12: Scattering Network Summary Diagram

5.6 Examples

In the examples below we did not apply any operation (nor logarithm of renormalization nor else). The sub-plots are from top to bottom the signal, and the L1, L2, L3 from the scattering. Ordinates are the j index.



Figure 13: Signal, L1, L2, L3 of Inia Perou (slow clicks) T1:4 Q1:32 J1:2 T2:256 Q2:1 J2:1 T3:16 Q3:1 J3:1



Figure 14: Signal, L1, L2, L3, Inia Perou (fast clicks) T1:4 Q1:20 J1:1 T2:128 Q2:1 J2:1 T3:2 Q3:4 J3:1

Some of the discontinuities seen on the y axis of L3 are from the way the results are stored and are due to the 2D representation of a 3D matrix. In fact here the L3 output is not of dimension 4 since L2 doesn't add a dimension because it is made of only one high-pass filter. Using only one high-pass filter for L3 would mean that the output of L3 is again of dimension 2.



Figure 15: Signal, L1, L2, L3, Bird (BIRDLIFE CLEF Challenge) T1:4 Q1:25 J1:4 T2:256 Q2:1 J2:1 T3:128 Q3:1 J3:1

6 Data representation

In order to appreciate the decompositions done either with spectrograms or the scattering network, this toolbox provides a imagesc like utility. The presentation of it is as follow : description of the import method, optimization of the rendering algorithm, representation of the data through OpenGL. Note that a special executable file is done and can be used on every .txt file containing a 2D matrix by calling it. More descriptions are in the annexA.

6.1 Import data

The Data is imported by reading a .txt file. This method doesn't need to know the size of the matrix then is not the most efficient one. When reading the file, the first line is used to learn the size of the matrix resulting for the other lines in a faster loading method. In fact, the first line uses the vector push back method that expands the size of the vector by 2 times its actual size when it's full meaning that for N points only $\log(N)$ expansions are made if we started from a vector of length 1. The following lines are directly loaded into a vector of the right size.

6.2 Optimization

Rendering is done using openGL. The matrix is decomposed into squares with nodes being representation of the matrix points. Using squares instead of triangles is more efficient in this particular case since all the points lie in a 2D plane. The values of the points are described through the colors. Another optimization used is through arrays (either vertex arrays, color arrays,...) by enabling client states. In fact passing directly the vertex, color and points arrays reduces the number of function calls and improve performances.

6.3 Colormap

To render good spectrograms or scalograms it is necessary to have a good colormap which is simply a $\mathbb{R} \to \mathbb{R}^3$ function mapping the value of a point to a RGB color. The one used here is the same as Matlab or Python.



Figure 16: Colormap used in the toolbox

This colormap need normalized data as input but offers very pleasant rendering colors.

6.4 Implementation

Given a matrix $N \times M$, squares are computed using or not the original aspect ratio. If the aspect ratio is kept then a matrix of size $N \times 2N$ for example will be displayed into $[0, 1] \times [0, 0.5]$ and this is true for every possible ratio. However this kind of displaying can become very unpleasant for extrem ratios (few rows and millions of columns for example). That is why it is possible to disregard the original ratio drawing the matrix into the unitary square by adding a simple argument when calling the function as follow:

```
./imshow myspectro.txt//keeping original ratio ./imshow 0 myspectro.txt//squared ratio % f(x)=0
```

The 0 option specifies to not keep the original aspect ratio meaning that now the image will be displayed as a square. During the data visualization it is possible to zoom in or out using + and -. Moves are possible through the mouse or the a,w,s,d keys.

7 Validation on a large scale bioacoustic data set

In order to test the whole tool on a real data set, we used the Bird Challenge 2014 3 contening nearly 100 hours of recording distributed on 14K files of bird songs from the Amazonian forest.

7.1 Recall of previous attempts

Previous attempts to run scattering decomposition on this dataset were difficult. We tried first[2] to run scatnet on each file, over time windows of 300 ms. The main difficulties were then the duration of the processing, and the cuts between the windows that was difficult to manage for a second stage of classification over the whole signal. In fact, the scatnet signal length limit is of 2^{15} bins.

7.2 Results with CIGAL

CIGAL allowed on 1 CPU 2600 Mhz of the UTLN server to run all the files in 72 hours for layer 0,1 and 2. Note that once layer 0 done, computation time for the other layers decreases significantly. We then get continuous decomposition as illustrated below for several files using Morlet wavelets.

7.3 Perspective on BIRD2015

We give in fig 17 to 20 the spectrogram versus the scattering of some tropical bird species from the Bird lifeclef challenge 2014 [ref]. They demonstrate the clarity of the features. We currently use these features to train Convolutional Neural Net for bird species identification.



Figure 17: XC73908.wav with T1:4 Q1:20 J1: 5, T2:256 Q2:1 J2:1, T3:128 Q3:1 J3:1, Spectrogram : 128 bins and 50% overlap

 $^{^3 {\}tt SABIOD.ORG}$



Figure 18: XC83327.wav with T1:4 Q1:20 J1: 5, T2:256 Q2:1 J2:1, T3:128 Q3:1 J3:1, Spectrogram : 128 bins and 50% overlap



Figure 19: XC81545.wav with T1:4 Q1:20 J1: 5, T2:256 Q2:1 J2:1, T3:128 Q3:1 J3:1, Spectrogram : 128 bins and 50% overlap



Figure 20: XC80923.wav with T1:4 Q1:20 J1: 5, T2:256 Q2:1 J2:1, T3:128 Q3:1 J3:1, Spectrogram : 128 bins and 50% overlap

8 Future work

Concerning computation time for the Scattering Network, here is a plot for each signal length (power of 2 from 2^8 to 2^{16}) with two layers. The asymptotic complexity is $O(n \log(n))$ for the scattering layer although here with the chosen coefficients we can't yet see the curve.



Figure 21: First Layer Coefficients : T=2,Q=8,J=8. Second Layer Coefficients :T=16,Q=1,J=1

With this toolbox, it will be possible to perform deep analysis on the massive

signal datasets taking advantage of the speed benefits of this implementation. More and more challenges are available such as the BIRDLIFE CLEF 2014 challenge⁴ offering a huge dataset.

This toolbox shall open new investigations into large scale databases, with complex and not yet well known sources, including multipath propagation. It may be useful for bioacoustic tropical forest databases or cetacean real time survey. The FFT3D algorithm will be added to this toolbox having the advantage of containing phse information which seems to be an interesting information for good representation.

Extension to GPU processing is one of the possible future work which could fully exploit the architecture of the presented algorithms.

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A README

Description of the folders :

- txt is where the results of the executable programs should be stored
- source contains all the algorithms and utilities
- exec contains code that use algorithms from source in order to make a usable command line executable for
 - scattering
 - spectrogram
 - imshow
- wav is a folder where the signals are stored, in this version some samples are already present for testing

The Makefile is used without any arguments, it will create all the executables (3) which can then be used as follow

SPECTROGRAM:

default arguments ./spectrogram wav/slow.wav //this will generate the file ./txt/spectrogram.txt which can be displayed as follow specify the window size ./spectrogram wav/slow.wav 256 specify the window size plus the overlapping

⁴http://sabiod.univ-tln.fr/public_data/BIRD_CHALLENGE/

./spectrogram wav/slow.wav 256 0.2

IMSHOW: keeping original ratio ./imshow txt/spectrogram.txt this utility acts like imagesc in matlab and can be used on any .txt storing a plain matrix square ratio ./imshow 0 txt/spectrogram.txt

SCATTERING:

arguments are in order : wav file, T1, Q1, J1, T2, Q2, J2,... the number of layers is determined by the number of args given. Doing the following command will save the layers in txt files into the txt directory ./scattering 8 20 4 8 1 1 32 1 1 wav/processed.wav

DISPLAYING THE SCATTERING

In order to display the wav plus all the scattering layers, a simple python file is used for convenience and is used by simply calling it and passing the wav on which the scattering has been done

python exec/Plots.py wav/processed.wav

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